## Exam 2 Solution

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## Problem 1



Let's use superposition. For $v_{o 1}=-4 v_{a}$, output because of $v_{a}$.

$$
\begin{array}{r}
v_{o 1} / v_{a}=-R_{1} / R_{2}=-4 \\
R_{2}=R_{1} / 4=250 \Omega
\end{array}
$$

Now, output because of positive terminal voltage, $v_{+}$is, $v_{o 2}=5 \mathrm{~V}$.

$$
\begin{gathered}
v_{o 2} / v_{+}=5=1+R_{2} / R_{1}=5 \\
v_{+}=1 \mathrm{~V} \\
v_{+}=\frac{R_{4}}{R_{3}+R_{4}} V_{1} \\
R_{3}=9 R_{4}
\end{gathered}
$$

$$
v_{+}=1 V
$$

We can choose $R_{4}$ to be $1 \mathrm{k} \Omega$ and $R_{3}$ to be $9 \mathrm{k} \Omega$, but other values are acceptable keeping the ratio same.

## Problem 2

The voltage in the capacitor drops while the plane operates, but the current remains constant. We know,

$$
i_{C_{s}}=C \frac{d v_{C_{s}}}{d t}
$$

So, if current is constant the voltage drops linearly across, $C_{s}$. Here, initial voltage across $C_{s}$ is $V_{1}=8 \mathrm{~V}$ and final voltage, $V_{2}=5 \mathrm{~V}$. So, energy delivered,

$$
E=\frac{V_{1}+V_{2}}{2} I t
$$

Also, energy delivered can be calculated using the difference of energy stored in the capacitor initially and finally.

$$
E=\frac{1}{2}\left(C_{s} V_{1}^{2}-C_{s} V_{2}^{2}\right) .
$$

So,

$$
\begin{array}{r}
\frac{V_{1}+V_{2}}{2} I t=\frac{1}{2}\left(C_{s} V_{1}^{2}-C_{s} V_{2}^{2}\right) \\
(8+5) \cdot 100 \cdot 10^{-3} \cdot 10 \cdot 60=C_{s}\left(8^{2}-5^{2}\right) \\
C_{s}=20 \mathrm{~F}
\end{array}
$$

## Problem 3

Both of the waveforms have same period, $T=3 \mathrm{~s}$. For the sinusoidal waveform, $V_{r m s, \sin }=V_{m} / \sqrt{2}=2 / \sqrt{2}=\sqrt{2} \mathrm{~V}$.

For the triangular wave we have to calculate the rms value.

$$
V_{r m s}=\sqrt{\frac{1}{T} \int_{0}^{T} v^{2}(t) d t}
$$

Now, the triangular wave is antisymmetric and if we square it will become symmetric. ON top of that, if we divide the whole period of the triangular waveshapes into four equal segments then the squared triangular wave will have same area under each fraction. So,

$$
V_{r m s, t r i}=\sqrt{\frac{4}{T} \int_{0}^{T / 4} v^{2}(t) d t}=\sqrt{\frac{4}{T} \int_{0}^{T / 4}\left(\frac{8 t}{3}\right)^{2} d t}=\frac{2}{\sqrt{3}}
$$

So,

$$
P_{r m s, s i n} / P_{r m s, t r i}=\left(P_{r m s, s i n} / P_{r m s, t r i}\right)^{2}=\left(\sqrt{(2) / \frac{2}{\sqrt{3}}}\right)^{2}=3 / 2
$$

## Problem 4

Piecewise linear version of the following plot


## Problem 5

(a) Here, $\mathbf{I}_{\mathrm{m}}=10 \angle 0^{\circ}$ and $\mathbf{Z}_{1}=30 \angle-40^{\circ}$. So,

$$
\begin{array}{r}
\mathbf{V}_{\mathbf{m}}=\mathbf{\mathbf { I Z } _ { \mathbf { 1 } }}=300 \angle-40^{\circ} \\
P=0.5 V_{m} I_{m} \cos \left(\theta_{v}-\theta_{i}\right) \\
P=0.5 \cdot 10 \cdot 300 \cdot \cos \left(-40^{\circ}-0^{\circ}\right) \\
P \approx 1149 \mathrm{~W}
\end{array}
$$

(b) For power factor to be unity, the equivalent impedance angle of the series combination needs to be $0{ }^{\circ}$. So, the imaginary part of the equivalent impedance needs to be 0 . Now, the imaginary part of $\mathbf{Z}_{1}$ is $30 \sin \left(-40^{\circ}\right)=-j 19.28 \Omega$. So, $\mathbf{Z}_{\mathbf{2}}$ can be any impedance with imaginary part $19.28 \Omega$.

## Problem 6

(a) at $\omega=0, C_{1}$ is open and $v_{+}=0 \mathrm{~V}$. So, $v_{o}=0$.

$$
v_{2} / v_{1}=0
$$

(b) at $\omega \rightarrow \infty, C_{1}$ is short and $v_{+}=v_{1} \mathrm{~V}$. However, $C_{2}$ is also short, so, $v_{2}=0$.

$$
v_{2} / v_{1}=0
$$

(c)

$$
H(j \omega)=\frac{V_{2}(j \omega)}{V_{1}(j \omega)}=\frac{R_{1}}{R_{1}+\frac{1}{j \omega C_{1}}} \frac{\frac{1}{j \omega C_{2}}}{R_{2}+\frac{1}{j \omega C_{2}}}=\frac{j \omega R_{1} C_{1}}{\left(1+j \omega R_{1} C_{1}\right)\left(1+j \omega R_{2} C_{2}\right)}
$$

(d) In part (c) if $\omega=0$, the magnitude of $H$ is obviously 0 .

If $\omega \rightarrow \infty, j \omega R_{1} C_{1}$ and $j \omega R_{2} C_{2}$ are much greater than 1 . So,

$$
H(j \omega)=\frac{j \omega R_{1} C_{1}}{j \omega R_{1} C_{1} \cdot j \omega R_{2} C_{2}}=\frac{1}{j \omega R_{2} C_{2}} \approx 0
$$

