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## UNIVERSITY OF CALIFORNIA College of Engineering Department of Electrical Engineering and Computer Sciences

## **B. E. BOSER**

## Midterm 2 October 26, 2004

EECS 105 FALL 2004

- Closed book, closed notes.
- No calculators.
- Copy your answers into marked boxes on exam sheets.
- Simplify numerical and algebraic results as much as possible. Up to 5 points penalty for results that are not reasonably simplified.
- Mark your name and SID at the top of the exam and all extra sheets.
- Be kind to the graders and write legibly. No credit for illegible results.

Problem 1 [25 points]



- $\begin{array}{lll} \underline{Given:} & \mu_n C_{ox} = 200 \mu A/V^2, \, V_{TN} = 1V, \, \lambda_n = 0.01 V^{-1} @ \ L = 1 \mu m \\ & g_m r_o >> 1 \\ & \ The \ circuit \ is \ biased \ such \ that \ all \ transistors \ are \ in \ saturation. \end{array}$
- a) [10 points] Find numerical values (not expressions) for:

$$I_{D3} = 1000$$
 µA

 $g_{m3} = 10000$ 

μS

$$I_{DS3} = I_{DS2} = \frac{\left(\frac{W}{L}\right)_2}{\left(\frac{W}{L}\right)_1} \cdot I_{DS1} = \frac{\left(\frac{100}{1}\right)}{\left(\frac{10}{1}\right)} \cdot 100\,\mu A = 1000\,\mu A$$
$$g_{m3} \approx \sqrt{\mu_n C_{ox}} \cdot \left(\frac{W}{L}\right)_3 \cdot 2 \cdot I_{DS3} = \sqrt{\left(200\frac{\mu A}{V^2}\right) \cdot \left(\frac{250}{1}\right) \cdot 2 \cdot 1000\,\mu A} = 10mS = 10000\,\mu S$$

b) [15 points] Find an *algebraic* expression for the small signal output resistance (at terminal  $v_{OUT}$ ) of the circuit as a function of transistor small-signal parameters. Use  $g_m r_o >> 1$  to simplify your result as much as possible:

 $r_{out} = 1/(g_{m3} + g_{mb3})$ 

$$r_{out} = r_{o2} // \frac{r_{o3}}{1 + (g_{m3} + g_{mb3}) \cdot r_{o3}} \approx r_{o2} // \frac{1}{(g_{m3} + g_{mb3})} \approx \frac{1}{(g_{m3} + g_{mb3})}$$

Problem 2 [25 points]



The above sketch shows a rough approximation of the electron drift velocity versus the electrical field in Silicon. For an NMOS transistor with L=0.1 $\mu$ m, W=10 $\mu$ m, and C<sub>ox</sub>=5fF/ $\mu$ m<sup>2</sup> calculate the following:

a) [10 points] What is the minimum  $V_{DS}$  (*numerical* value) for which current flow is limited by the thermal carrier drift velocity? Assume that the field in the channel is uniform.

$$V_{\rm DS} = 0.1 \qquad \qquad V$$

$$E_{critical} = \frac{1V}{1\mu m} = \frac{V_{DS}}{L} = \frac{V_{DS}}{0.1\mu m} \Longrightarrow V_{DS} = 0.1V$$

b) [15 points] Find the *numerical* value of the maximum drain current  $I_D$  for  $V_{GS}$ - $V_{TH}=1V$ . Hint: get the current from the channel charge and its velocity.



Average channel charge = 
$$\frac{(Ch \arg e@ source) + (Ch \arg e@ drain)}{2}$$

$$= \frac{C_{ox} \cdot W \cdot L \cdot [(V_{GS} - V_{TH}) + (V_{GD} - V_{TH})]}{2} = \frac{C_{ox} \cdot W \cdot L \cdot [(V_{GS} - V_{TH}) + (V_{GS} - V_{DS} - V_{TH})]}{2}$$

$$= \frac{5 \frac{fF}{\mu m^2} \cdot 10 \mu m \cdot 0.1 \mu m \cdot [1V + 0.9V]}{2} = 4.75 fC$$

$$\therefore Average channel charge per unit length = \frac{Average channel charge}{L} = 47.5 \frac{fC}{\mu m}$$

$$I_D = (Average channel charge per unit length) \cdot (velocity)$$

: 
$$I_D = 47.5 \frac{fC}{\mu m} \cdot 10^{11} \frac{\mu m}{s} = 4.75 mA$$

## Problem 3 [25 points]



The circuit shown above is biased so that all transistors are in saturation. Draw a small signal model (label all elements with appropriate symbols, e.g.  $g_{m1}$ ,  $r_{o2}$ ) and find an *algebraic* expression for the small-signal voltage gain  $a_v = v_{out}/v_{in}$  as a function of small-signal parameters ( $g_m$ 's and  $r_o$ 's). Use  $g_m r_o >> 1$  to simplify your result.

Small-signal model (neatness counts) [13 points]:



 $a_v = -\mathbf{g_{m1}}/\mathbf{g_{m2}}$ 

$$a_v = -g_{m1} \cdot (r_{o1} // r_{o2} // \frac{1}{g_{m2}}) \approx \frac{-g_{m1}}{g_{m2}}$$

Problem 4 [23 points]



The circuit shown above is biased so that the transistor is in saturation.

a) [8 points] What is the type of this amplifier?

Common	<u>Gate</u>			

b) [15 points] Find an *algebraic* expression for the small-signal voltage ratio  $v_2/v_1$  for  $i_s=0$  as a function of R<sub>1</sub>, R<sub>2</sub>, and transistor small-signal parameters. Hint: you may find small-signal model very helpful to answer this question.

 $v_2/v_1 = \frac{(1+g_m r_o)}{r_o + R_2} \cdot R_2$